

Quad Low Power Operational Amplifiers

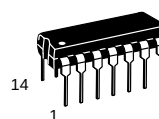
The LM324 series are low-cost, quad operational amplifiers with true differential inputs. They have several distinct advantages over standard operational amplifier types in single supply applications. The quad amplifier can operate at supply voltages as low as 3.0 V or as high as 32 V with quiescent currents about one-fifth of those associated with the MC1741 (on a per amplifier basis). The common mode input range includes the negative supply, thereby eliminating the necessity for external biasing components in many applications. The output voltage range also includes the negative power supply voltage.

- Short Circuited Protected Outputs
- True Differential Input Stage
- Single Supply Operation: 3.0 V to 32 V
- Low Input Bias Currents: 100 nA Maximum (LM324A)
- Four Amplifiers Per Package
- Internally Compensated
- Common Mode Range Extends to Negative Supply
- Industry Standard Pinouts
- ESD Clamps on the Inputs Increase Ruggedness without Affecting Device Operation

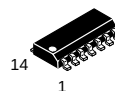
LM324, LM324A, LM224, LM2902, LM2902V

QUAD DIFFERENTIAL INPUT OPERATIONAL AMPLIFIERS

SEMICONDUCTOR TECHNICAL DATA



N SUFFIX
PLASTIC PACKAGE
CASE 646
(LM224, LM324,
LM2902 Only)



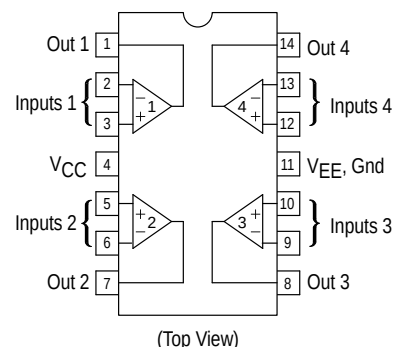
D SUFFIX
PLASTIC PACKAGE
CASE 751A
(SO-14)

MAXIMUM RATINGS ($T_A = +25^\circ\text{C}$, unless otherwise noted.)

Rating	Symbol	LM224 LM324, LM324A	LM2902, LM2902V	Unit
Power Supply Voltages Single Supply Split Supplies	V_{CC} V_{CC}, V_{EE}	32 ± 16	26 ± 13	Vdc
Input Differential Voltage Range (See Note 1)	V_{IDR}	± 32	± 26	Vdc
Input Common Mode Voltage Range	V_{ICR}	-0.3 to 32	-0.3 to 26	Vdc
Output Short Circuit Duration	t_{SC}	Continuous		
Junction Temperature	T_J	150		$^\circ\text{C}$
Storage Temperature Range	T_{stg}	-65 to +150		$^\circ\text{C}$
Operating Ambient Temperature Range	T_A	-25 to +85 0 to +70	-40 to +105 -40 to +125	$^\circ\text{C}$

NOTE: 1. Split Power Supplies.

PIN CONNECTIONS



ORDERING INFORMATION

Device	Operating Temperature Range	Package
LM2902D	$T_A = -40^\circ$ to $+105^\circ\text{C}$	SO-14
LM2902N		Plastic DIP
LM2902VD	$T_A = -40^\circ$ to $+125^\circ\text{C}$	SO-14
LM2902VN		Plastic DIP
LM224D	$T_A = -25^\circ$ to $+85^\circ\text{C}$	SO-14
LM224N		Plastic DIP
LM324AD	$T_A = 0^\circ$ to $+70^\circ\text{C}$	SO-14
LM324AN		Plastic DIP
LM324D		SO-14
LM324N		Plastic DIP

LM324, LM324A, LM224, LM2902, LM2902V

ELECTRICAL CHARACTERISTICS ($V_{CC} = 5.0\text{ V}$, $V_{EE} = \text{Gnd}$, $T_A = 25^\circ\text{C}$, unless otherwise noted.)

Characteristics	Symbol	LM224			LM324A			LM324			LM2902			LM2902V			Unit
		Min	Typ	Max	Min	Typ	Max	Min	Typ	Max	Min	Typ	Max	Min	Typ	Max	
Input Offset Voltage $V_{CC} = 5.0\text{ V}$ to 30 V (26 V for LM2902, V), $V_{ICR} = 0\text{ V}$ to $V_{CC} - 1.7\text{ V}$, $V_O = 1.4\text{ V}$, $R_S = 0\ \Omega$ $T_A = 25^\circ\text{C}$ $T_A = T_{\text{high}}^{(1)}$ $T_A = T_{\text{low}}^{(1)}$	V_{IO}	–	2.0	5.0	–	2.0	3.0	–	2.0	7.0	–	2.0	7.0	–	2.0	7.0	mV
		–	–	7.0	–	–	5.0	–	–	9.0	–	–	10	–	–	13	
		–	–	7.0	–	–	5.0	–	–	9.0	–	–	10	–	–	10	
Average Temperature Coefficient of Input Offset Voltage $T_A = T_{\text{high}}$ to $T_{\text{low}}^{(1)}$	$\Delta V_{IO}/\Delta T$	–	7.0	–	–	7.0	30	–	7.0	–	–	7.0	–	–	7.0	–	$\mu\text{V}/^\circ\text{C}$
Input Offset Current $T_A = T_{\text{high}}$ to $T_{\text{low}}^{(1)}$	I_{IO}	–	3.0	30	–	5.0	30	–	5.0	50	–	5.0	50	–	5.0	50	nA
		–	–	100	–	–	75	–	–	150	–	–	200	–	–	200	
Average Temperature Coefficient of Input Offset Current $T_A = T_{\text{high}}$ to $T_{\text{low}}^{(1)}$	$\Delta I_{IO}/\Delta T$	–	10	–	–	10	300	–	10	–	–	10	–	–	10	–	$\text{pA}/^\circ\text{C}$
Input Bias Current $T_A = T_{\text{high}}$ to $T_{\text{low}}^{(1)}$	I_{IB}	–	–90	–150	–	–45	–100	–	–90	–250	–	–90	–250	–	–90	–250	nA
		–	–	–300	–	–	–200	–	–	–500	–	–	–500	–	–	–500	
Input Common Mode Voltage Range ⁽²⁾ $V_{CC} = 30\text{ V}$ (26 V for LM2902, V) $V_{CC} = 30\text{ V}$ (26 V for LM2902, V), $T_A = T_{\text{high}}$ to T_{low}	V_{ICR}	0	–	28.3	0	–	28.3	0	–	28.3	0	–	24.3	0	–	24.3	V
		0	–	28	0	–	28	0	–	28	0	–	24	0	–	24	
Differential Input Voltage Range	V_{IDR}	–	–	V_{CC}	–	–	V_{CC}	–	–	V_{CC}	–	–	V_{CC}	–	–	V_{CC}	V
Large Signal Open Loop Voltage Gain $R_L = 2.0\text{ k}\Omega$, $V_{CC} = 15\text{ V}$, for Large V_O Swing, $T_A = T_{\text{high}}$ to $T_{\text{low}}^{(1)}$	A_{VOL}	50 25	100 –	– –	25 15	100 –	– –	25 15	100 –	– –	25 15	100 –	– –	25 15	100 –	– –	V/mV
Channel Separation $10\text{ kHz} \leq f \leq 20\text{ kHz}$, Input Referenced	CS	–	–120	–	–	–120	–	–	–120	–	–	–120	–	–	–120	–	dB
Common Mode Rejection, $R_S \leq 10\text{ k}\Omega$	CMR	70	85	–	65	70	–	65	70	–	50	70	–	50	70	–	dB
Power Supply Rejection	PSR	65	100	–	65	100	–	65	100	–	50	100	–	50	100	–	dB
Output Voltage – High Limit ($T_A = T_{\text{high}}$ to $T_{\text{low}}^{(1)}$) $V_{CC} = 5.0\text{ V}$, $R_L = 2.0\text{ k}\Omega$, $T_A = 25^\circ\text{C}$ $V_{CC} = 30\text{ V}$ (26 V for LM2902, V), $R_L = 2.0\text{ k}\Omega$ $V_{CC} = 30\text{ V}$ (26 V for LM2902, V), $R_L = 10\text{ k}\Omega$	V_{OH}	3.3 26 27	3.5 – 28	– – –	3.3 26 27	3.5 – 28	– – –	3.3 26 27	3.5 – 28	– – –	3.3 22 23	3.5 – 24	– – –	3.3 22 23	3.5 – 24	– – –	V

NOTES: 1. $T_{\text{low}} = -25^\circ\text{C}$ for LM224
 $= 0^\circ\text{C}$ for LM324, A
 $= -40^\circ\text{C}$ for LM2902
 $= -40^\circ\text{C}$ for LM2902V
 $T_{\text{high}} = +85^\circ\text{C}$ for LM224
 $= +70^\circ\text{C}$ for LM324, A
 $= +105^\circ\text{C}$ for LM2902
 $= +125^\circ\text{C}$ for LM2902V

2. The input common mode voltage or either input signal voltage should not be allowed to go negative by more than 0.3 V . The upper end of the common mode voltage range is $V_{CC} - 1.7\text{ V}$.

LM324, LM324A, LM224, LM2902, LM2902V

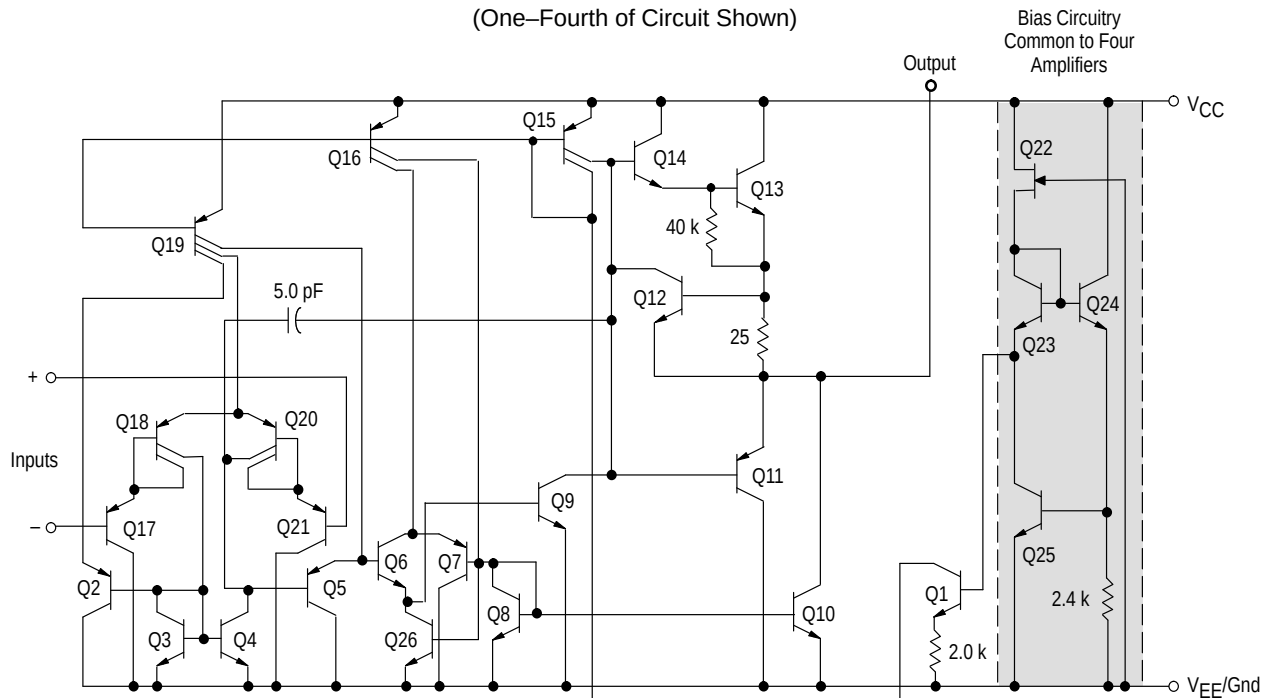
ELECTRICAL CHARACTERISTICS ($V_{CC} = 5.0\text{ V}$, $V_{EE} = \text{Gnd}$, $T_A = 25^\circ\text{C}$, unless otherwise noted.)

Characteristics	Symbol	LM224			LM324A			LM324			LM2902			LM2902V			Unit
		Min	Typ	Max	Min	Typ	Max	Min	Typ	Max	Min	Typ	Max	Min	Typ	Max	
Output Voltage – Low Limit, $V_{CC} = 5.0\text{ V}$, $R_L = 10\text{ k}\Omega$, $T_A = T_{\text{high}}$ to $T_{\text{low}}^{(1)}$	V_{OL}	–	5.0	20	–	5.0	20	–	5.0	20	–	5.0	100	–	5.0	100	mV
Output Source Current ($V_{ID} = +1.0\text{ V}$, $V_{CC} = 15\text{ V}$) $T_A = 25^\circ\text{C}$ $T_A = T_{\text{high}}$ to $T_{\text{low}}^{(1)}$	I_{O+}	20 10	40 20	– –	20 10	40 20	– –	20 10	40 20	– –	20 10	40 20	– –	20 10	40 20	– –	mA
Output Sink Current ($V_{ID} = -1.0\text{ V}$, $V_{CC} = 15\text{ V}$) $T_A = 25^\circ\text{C}$ $T_A = T_{\text{high}}$ to $T_{\text{low}}^{(1)}$ ($V_{ID} = -1.0\text{ V}$, $V_O = 200\text{ mV}$, $T_A = 25^\circ\text{C}$)	I_{O-}	10 5.0 12	20 8.0 50	– – –	10 5.0 12	20 8.0 50	– – –	10 5.0 12	20 8.0 50	– – –	10 5.0 –	20 8.0 –	– – –	10 5.0 –	20 8.0 –	– – –	mA μA
Output Short Circuit to Ground ⁽³⁾	I_{SC}	–	40	60	–	40	60	–	40	60	–	40	60	–	40	60	mA
Power Supply Current ($T_A = T_{\text{high}}$ to $T_{\text{low}}^{(1)}$) $V_{CC} = 30\text{ V}$ (26 V for LM2902, V), $V_O = 0\text{ V}$, $R_L = \infty$ $V_{CC} = 5.0\text{ V}$, $V_O = 0\text{ V}$, $R_L = \infty$	I_{CC}	– –	– –	3.0 1.2	– –	1.4 0.7	3.0 1.2	– –	– –	3.0 1.2	– –	– –	3.0 1.2	– –	– –	3.0 1.2	mA

NOTES: 1. $T_{\text{low}} = -25^\circ\text{C}$ for LM224
 $= 0^\circ\text{C}$ for LM324, A
 $= -40^\circ\text{C}$ for LM2902
 $= -40^\circ\text{C}$ for LM2902V
 $T_{\text{high}} = +85^\circ\text{C}$ for LM224
 $= +70^\circ\text{C}$ for LM324, A
 $= +105^\circ\text{C}$ for LM2902
 $= +125^\circ\text{C}$ for LM2902V

2. The input common mode voltage or either input signal voltage should not be allowed to go negative by more than 0.3 V. The upper end of the common mode voltage range is $V_{CC} - 1.7\text{ V}$.

Representative Circuit Diagram
(One-Fourth of Circuit Shown)

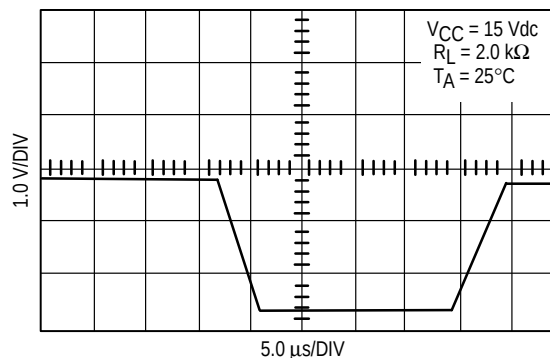


LM324, LM324A, LM224, LM2902, LM2902V

CIRCUIT DESCRIPTION

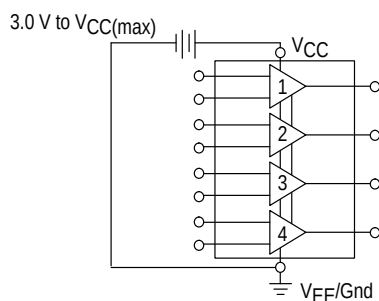
The LM324 series is made using four internally compensated, two-stage operational amplifiers. The first stage of each consists of differential input devices Q20 and Q18 with input buffer transistors Q21 and Q17 and the differential to single ended converter Q3 and Q4. The first stage performs not only the first stage gain function but also performs the level shifting and transconductance reduction functions. By reducing the transconductance, a smaller compensation capacitor (only 5.0 pF) can be employed, thus saving chip area. The transconductance reduction is accomplished by splitting the collectors of Q20 and Q18. Another feature of this input stage is that the input common mode range can include the negative supply or ground, in single supply operation, without saturating either the input devices or the differential to single-ended converter. The second stage consists of a standard current source load amplifier stage.

Large Signal Voltage Follower Response

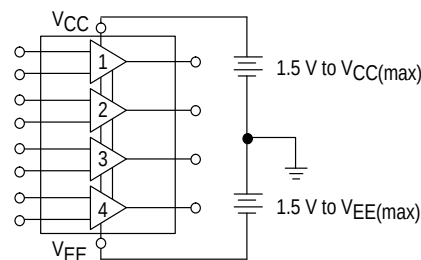


Each amplifier is biased from an internal-voltage regulator which has a low temperature coefficient thus giving each amplifier good temperature characteristics as well as excellent power supply rejection.

Single Supply



Split Supplies



LM324, LM324A, LM224, LM2902, LM2902V

Figure 1. Input Voltage Range

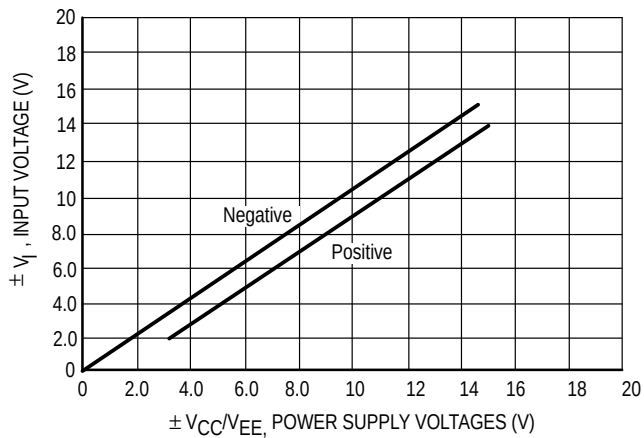


Figure 2. Open Loop Frequency

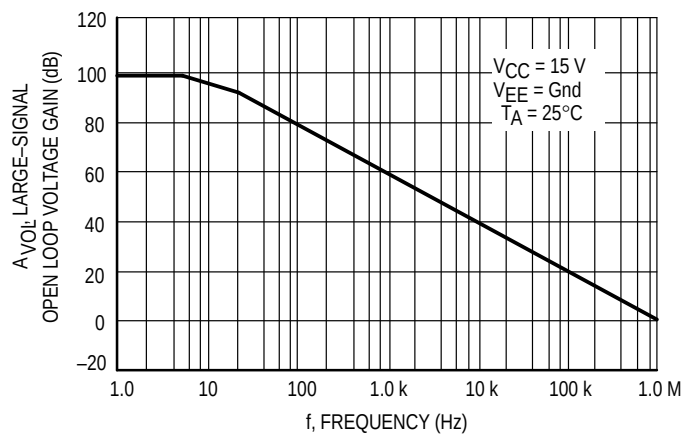


Figure 3. Large-Signal Frequency Response

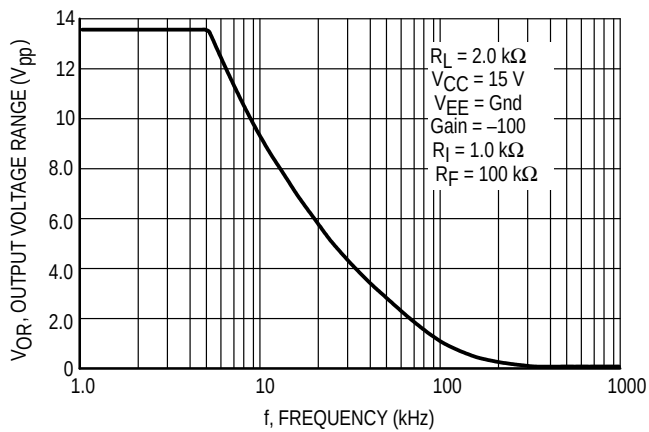


Figure 4. Small-Signal Voltage Follower Pulse Response (Noninverting)

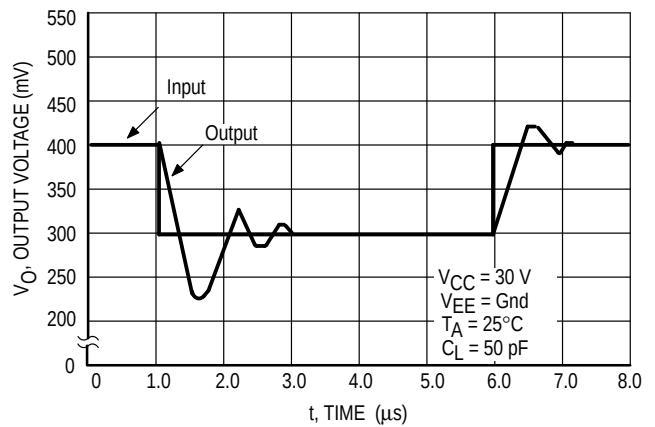


Figure 5. Power Supply Current versus Power Supply Voltage

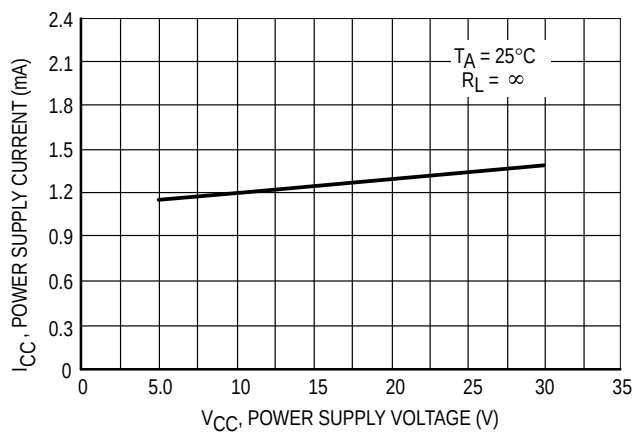


Figure 6. Input Bias Current versus Power Supply Voltage

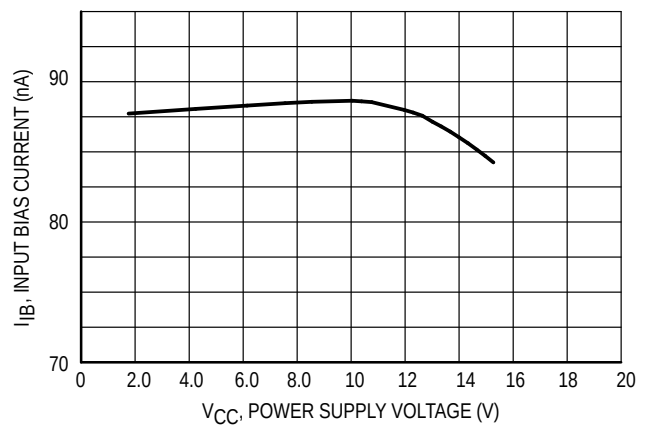


Figure 7. Voltage Reference

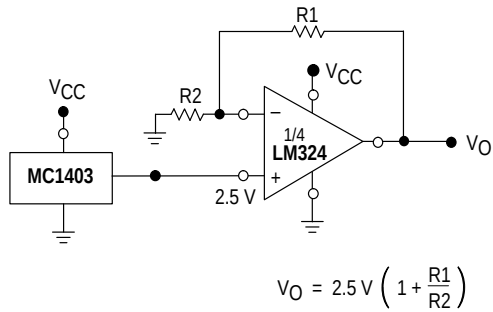


Figure 8. Wien Bridge Oscillator

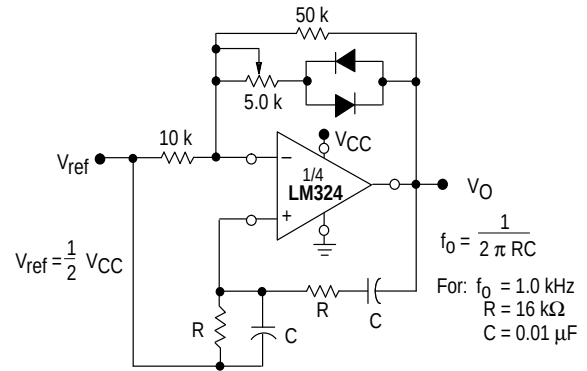


Figure 9. High Impedance Differential Amplifier

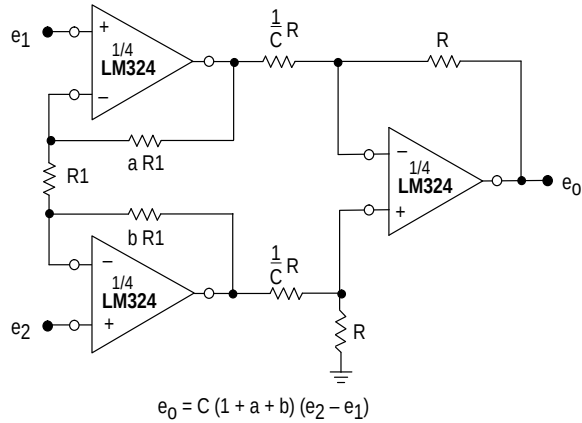


Figure 10. Comparator with Hysteresis

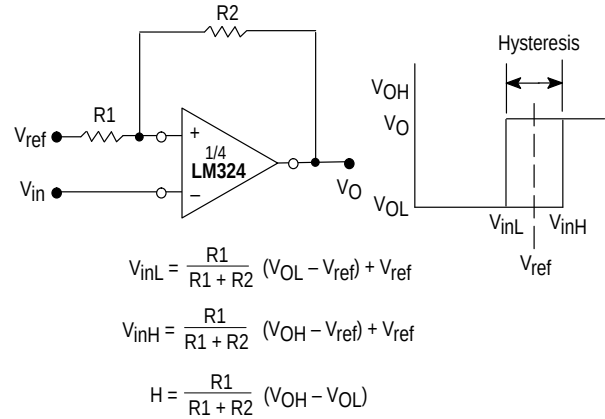


Figure 11. Bi-Quad Filter

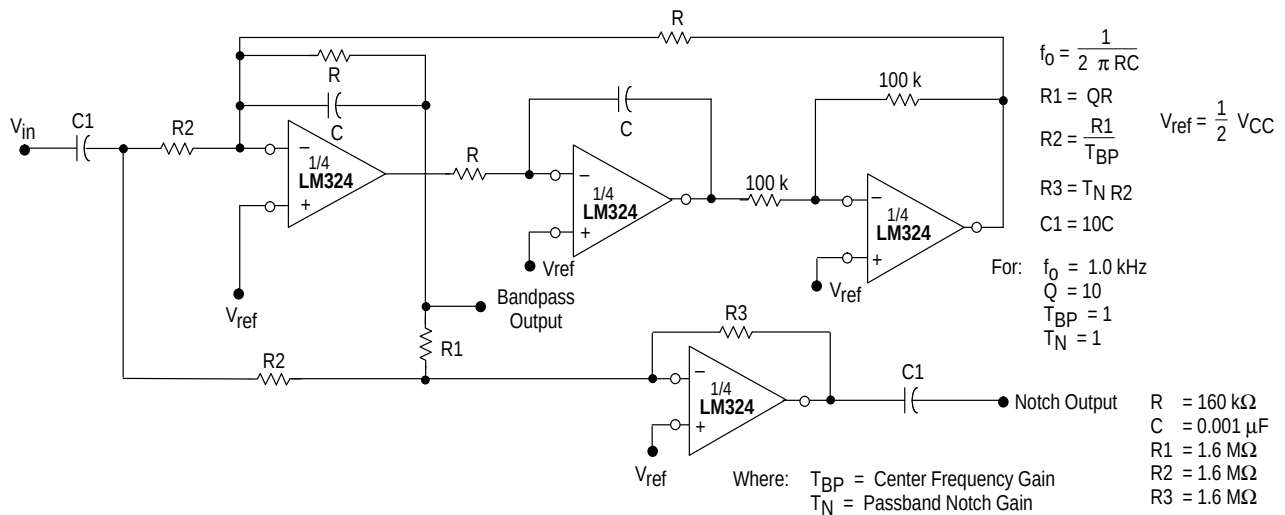


Figure 12. Function Generator

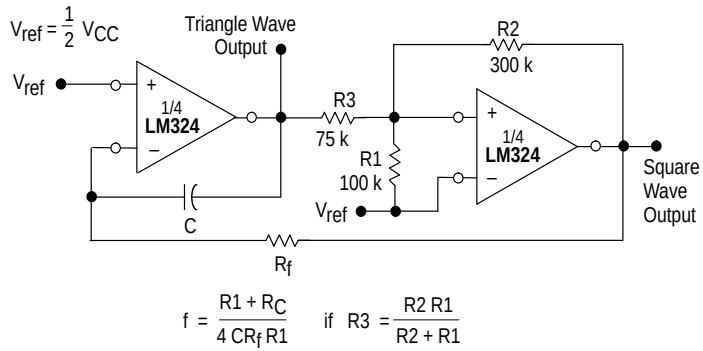
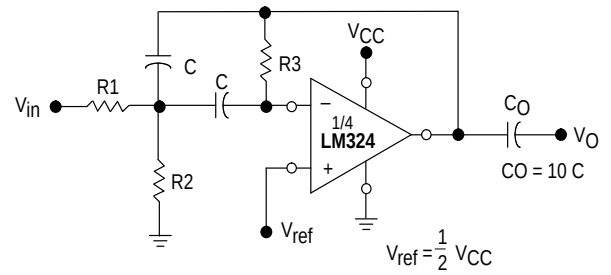


Figure 13. Multiple Feedback Bandpass Filter



Given: f_0 = center frequency
 $A(f_0)$ = gain at center frequency

Choose value f_0 , C

Then: $R_3 = \frac{Q}{\pi f_0 C}$

$R_1 = \frac{R_3}{2 A(f_0)}$

$R_2 = \frac{R_1 R_3}{4 Q^2 R_1 - R_3}$

For less than 10% error from operational amplifier, $\frac{Q_0 f_0}{BW} < 0.1$

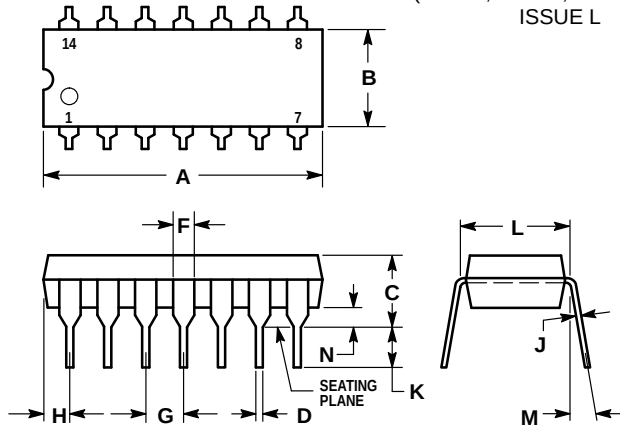
where f_0 and BW are expressed in Hz.

If source impedance varies, filter may be preceded with voltage follower buffer to stabilize filter parameters.

LM324, LM324A, LM224, LM2902, LM2902V

OUTLINE DIMENSIONS

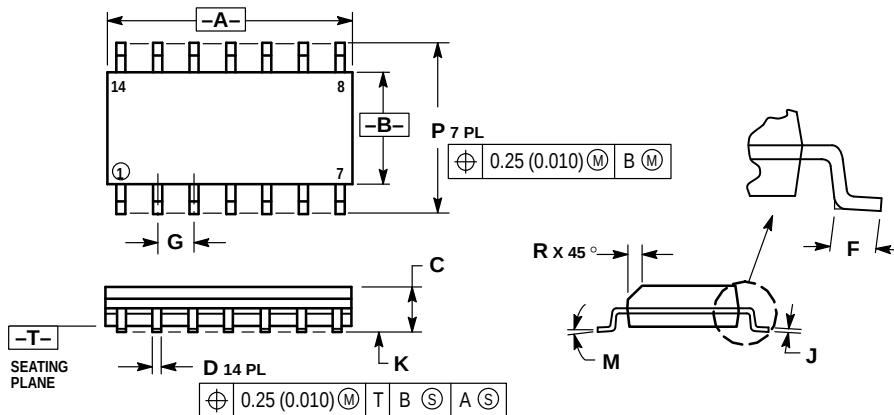
N SUFFIX PLASTIC PACKAGE CASE 646-06 (LM224, LM324, LM2902 Only) ISSUE L



- NOTES:
1. LEADS WITHIN 0.13 (0.005) RADIUS OF TRUE POSITION AT SEATING PLANE AT MAXIMUM MATERIAL CONDITION.
 2. DIMENSION L TO CENTER OF LEADS WHEN FORMED PARALLEL.
 3. DIMENSION B DOES NOT INCLUDE MOLD FLASH.
 4. ROUNDED CORNERS OPTIONAL.

DIM	INCHES		MILLIMETERS	
	MIN	MAX	MIN	MAX
A	0.715	0.770	18.16	19.56
B	0.240	0.260	6.10	6.60
C	0.145	0.185	3.69	4.69
D	0.015	0.021	0.38	0.53
F	0.040	0.070	1.02	1.78
G	0.100 BSC		2.54 BSC	
H	0.052	0.095	1.32	2.41
J	0.008	0.015	0.20	0.38
K	0.115	0.135	2.92	3.43
L	0.300 BSC		7.62 BSC	
M	0°	10°	0°	10°
N	0.015	0.039	0.39	1.01

D SUFFIX PLASTIC PACKAGE CASE 751A-03 (SO-14) ISSUE F



- NOTES:
1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
 2. CONTROLLING DIMENSION: MILLIMETER.
 3. DIMENSIONS A AND B DO NOT INCLUDE MOLD PROTRUSION.
 4. MAXIMUM MOLD PROTRUSION 0.15 (0.006) PER SIDE.
 5. DIMENSION D DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.127 (0.005) TOTAL IN EXCESS OF THE D DIMENSION AT MAXIMUM MATERIAL CONDITION.

DIM	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
A	8.55	8.75	0.337	0.344
B	3.80	4.00	0.150	0.157
C	1.35	1.75	0.054	0.068
D	0.35	0.49	0.014	0.019
F	0.40	1.25	0.016	0.049
G	1.27 BSC		0.050 BSC	
J	0.19	0.25	0.008	0.009
K	0.10	0.25	0.004	0.009
M	0°	7°	0°	7°
P	5.80	6.20	0.228	0.244
R	0.25	0.50	0.010	0.019

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