

DATA SHEET

TDA7057AQ

**2 × 8 W stereo BTL audio output
amplifier with DC volume control**

Product specification
Supersedes data of 1997 July 15
File under Integrated Circuits, IC01

1998 Apr 07

2 × 8 W stereo BTL audio output amplifier with DC volume control

TDA7057AQ

FEATURES

- DC volume control
- Few external components
- Mute mode
- Thermal protection
- Short-circuit proof
- No switch-on and switch-off clicks
- Good overall stability
- Low power consumption
- Low HF radiation
- ESD protected on all pins.

GENERAL DESCRIPTION

The TDA7057AQ is a stereo BTL output amplifier with DC volume control. The device is designed for use in TVs and monitors, but is also suitable for battery-fed portable recorders and radios.

Missing Current Limiter (MCL)

A MCL protection circuit is built-in. The MCL circuit is activated when the difference in current between the output terminal of each amplifier exceeds 100 mA (typical 300 mA). This level of 100 mA allows for single-ended headphone applications.

QUICK REFERENCE DATA

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
V_P	supply voltage		4.5	–	18	V
P_{out}	output power	$V_P = 12\text{ V}; R_L = 16\ \Omega$	3.0	3.5	–	W
		$V_P = 12\text{ V}; R_L = 8\ \Omega$	–	5.3	–	W
		$V_P = 15\text{ V}; R_L = 8\ \Omega$	–	8	–	W
G_v	voltage gain		39.5	40.5	41.5	dB
ΔG_v	voltage gain control		68	73.5	–	dB
$I_{q(tot)}$	total quiescent current	$V_P = 12\text{ V}; R_L = \infty$	–	22	25	mA
THD	total harmonic distortion	$P_0 = 0.5\text{ W}$	–	0.3	1	%

ORDERING INFORMATION

TYPE NUMBER	PACKAGE		
	NAME	DESCRIPTION	VERSION
TDA7057AQ	DBS13P	plastic DIL-bent-SIL power package; 13 leads (lead length 12 mm)	SOT141-6

$2 \times 8 \text{ W}$ stereo BTL audio output amplifier with DC volume control

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BLOCK DIAGRAM

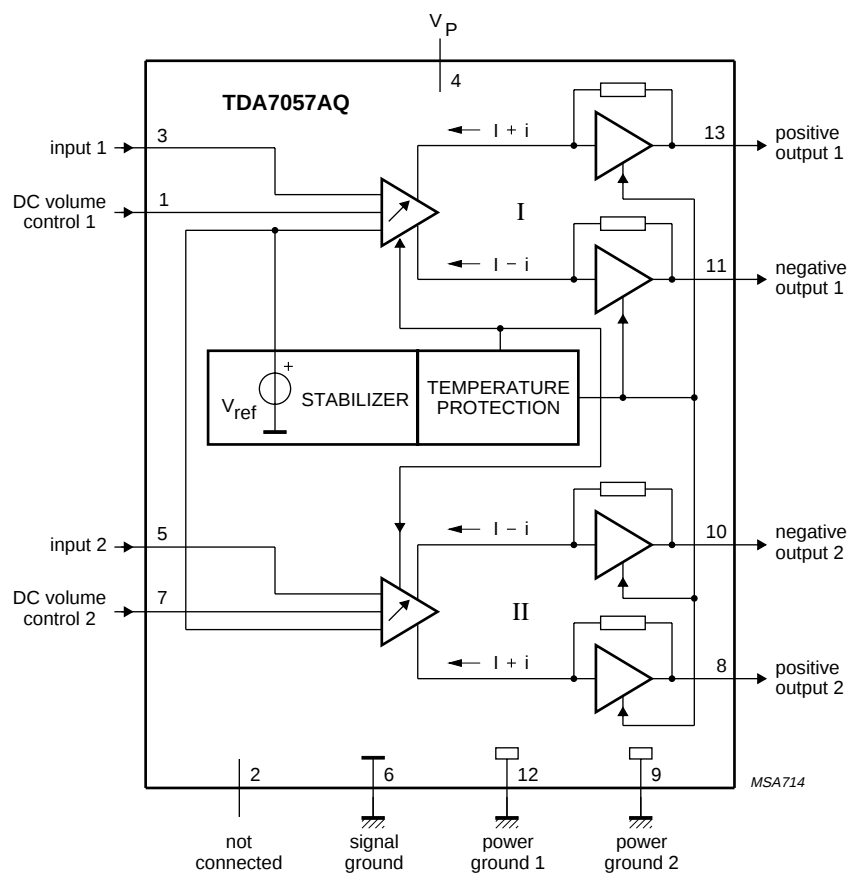


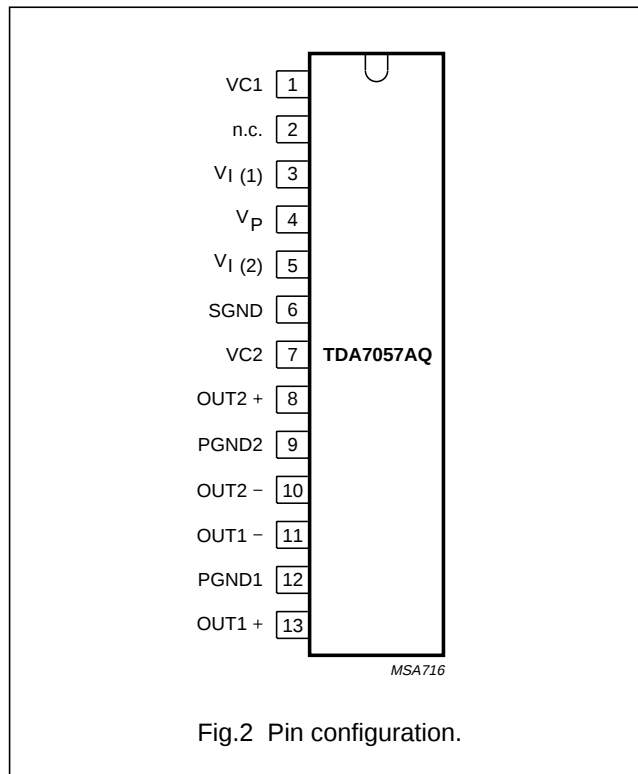
Fig.1 Block diagram.

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PINNING

SYMBOL	PIN	DESCRIPTION
VC1	1	DC volume control 1
n.c.	2	not connected
V _{I(1)}	3	voltage input 1
V _P	4	positive supply voltage
V _{I(2)}	5	voltage input 2
SGND	6	signal ground
VC2	7	DC volume control 2
OUT2+	8	positive output 2
PGND2	9	power ground 2
OUT2–	10	negative output 2
OUT1–	11	negative output 1
PGND1	12	power ground 1
OUT1+	13	positive output 1



FUNCTIONAL DESCRIPTION

The TDA7057AQ is a stereo output amplifier with two DC volume control stages. The device is designed for TVs and monitors, but is also suitable for battery-fed portable recorders and radios.

In conventional DC volume control circuits the control or input stage is AC-coupled to the output stage via external capacitors to keep the offset voltage low.

In the TDA7057AQ the two DC volume control stages are integrated into the input stages so that no coupling capacitors are required and a low offset voltage is still maintained. The minimum supply voltage also remains low.

The BTL principle offers the following advantages:

- Lower peak value of the supply current
- The frequency of the ripple on the supply voltage is twice the signal frequency.

Consequently, a reduced power supply with smaller capacitors can be used which results in cost reductions.

For portable applications there is a trend to decrease the supply voltage, resulting in a reduction of output power at conventional output stages. Using the BTL principle increases the output power.

The maximum gain of the amplifier is fixed at 40.5 dB. The DC volume control stages have a logarithmic control characteristic. Therefore, the total gain can be controlled from +40.5 dB to –33 dB. If the DC volume control voltage falls below 0.4 V, the device will switch to the mute mode.

The amplifier is a short-circuit protected to ground, V_P and across the load. A thermal protection circuit is also implemented. If the crystal temperature rises above +150 °C the gain will be reduced, thereby reducing the output power.

Special attention is given to switch-on and switch-off clicks, low HF radiation and a good overall stability.

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LIMITING VALUES

In accordance with the Absolute Maximum Rating System (IEC 134).

SYMBOL	PARAMETER	CONDITIONS	MIN.	MAX.	UNIT
V_P	supply voltage		–	18	V
I_{ORM}	repetitive peak output current		–	1.25	A
I_{OSM}	non-repetitive peak output current		–	1.5	A
P_{tot}	total power dissipation	$T_{case} < 60\text{ °C}$	–	22.5	W
T_{amb}	operating ambient temperature		–40	+85	°C
T_{stg}	storage temperature		–55	+150	°C
T_{vj}	virtual junction temperature		–	150	°C
t_{sc}	short-circuit time		–	1	hr
V_n	input voltage pins 1, 3, 5 and 7		–	5	V

THERMAL CHARACTERISTICS

SYMBOL	PARAMETER	VALUE	UNIT
$R_{th\ j-c}$	thermal resistance from junction to case	4	K/W
$R_{th\ j-a}$	thermal resistance from junction to ambient in free air	40	K/W

Power dissipation

Assume $V_P = 12\text{ V}$ and $R_L = 16\text{ }\Omega$. The maximum sine wave dissipation is $2 \times 1.8\text{ W} = 3.6\text{ W}$.

At $T_{amb(max)} = 60\text{ °C}$:

$$R_{th\ tot} = (150 - 60)/3.6 = 25\text{ K/W.}$$

$$R_{th\ tot} = R_{th\ j-c} + R_{th\ c-hs} + R_{th\ hs}$$

$$R_{th\ c-hs} + R_{th\ hs} = 25 - 4 = 21\text{ K/W.}$$

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CHARACTERISTICS

$V_P = 12\text{ V}$; $T_{\text{amb}} = 25\text{ °C}$; $f_i = 1\text{ kHz}$; $R_L = 16\text{ }\Omega$; unless otherwise specified (see Fig.13).

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
V _P	voltage supply		4.5	–	18.5	V
I _{q(tot)}	total quiescent current	V _P = 12 V; R _L = ∞; note 1	–	22	25	mA
Maximum gain; V _{1,7} ≥ 1.4 V						
P _O	output power	THD = 10%; R _L = 16 Ω	3.0	3.5	–	W
		THD = 10%; R _L = 8 Ω	–	5.3	–	W
		THD=10%; R _L = 8 Ω; V _P = 15 V	–	8	–	W
THD	total harmonic distortion	P _O = 0.5 W	–	0.3	1	%
G _V	voltage gain		39.5	40.5	41.5	dB
V _{i(rms)}	input signal handling (RMS value)	G _V = 0 dB; THD < 1%	1	–	–	V
V _{o(n)}	noise output voltage	f _i = 500 kHz; note 2	–	210	–	μV
B	bandwidth	at –1 dB	–	note 3	–	dB
SVRR	supply voltage ripple rejection	note 4	34	38	–	dB
V _{OS}	DC output offset voltage	V ₁₃ - V ₁₁ and V ₁₀ - V ₈	–	0	200	mV
Z _i	input impedence (pins 3 and 5)		15	20	25	kΩ
α _{CS}	channel separation	R _S = 5 kΩ	40	–	–	dB
G _V	channel unbalance	note 5	–	–	1	dB
		G ₁ = 0 dB; note 6	–	–	1	dB
Mute position; V ₁ = V ₇ =0.4 V ±30 mV						
V _{o(mute)}	output voltage in mute position	V _i = 1.0 V; note 7	–	35	45	μV
DC volume control						
ΔG _V	gain control range		68	73.5	–	dB
I _{DC}	volume control current	V ₁ = V ₇ = 0 V	–20	–25	–30	μA

Notes

1. With a load connected to the outputs the quiescent current will increase, the maximum value of this increase being equal to the DC output offset voltage divided by R_L .
2. The noise output voltage (RMS value) at $f_i = 500\text{ kHz}$ is measured with $R_S = 0\text{ }\Omega$ and bandwidth = 5 kHz.
3. 20 Hz to 300 kHz (typical).
4. The ripple rejection is measured with $R_S = 0\text{ }\Omega$ and $f = 100\text{ Hz}$ to 10 kHz. The ripple voltage ($V_{\text{ripple}} = 200\text{ mV RMS}$) is applied to the positive supply rail.
5. The channel unbalance is measured with $V_{DC1} = V_{DC2}$.
6. The channel unbalance at $G_1 = 0\text{ dB}$ is measured with $V_{DC1} = V_{DC2}$.
7. The noise output voltage (RMS value) is measured with $R_S = 5\text{ k}\Omega$ unweighted.

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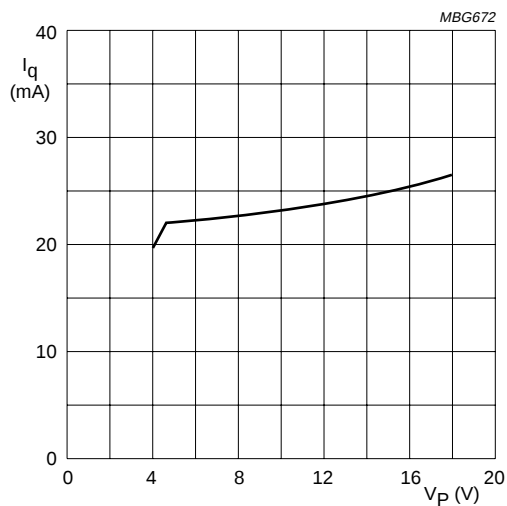
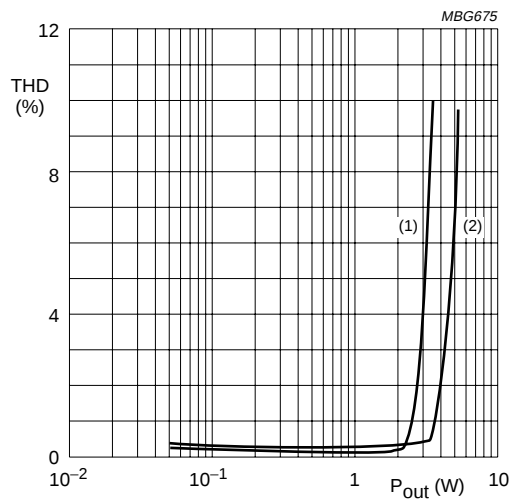
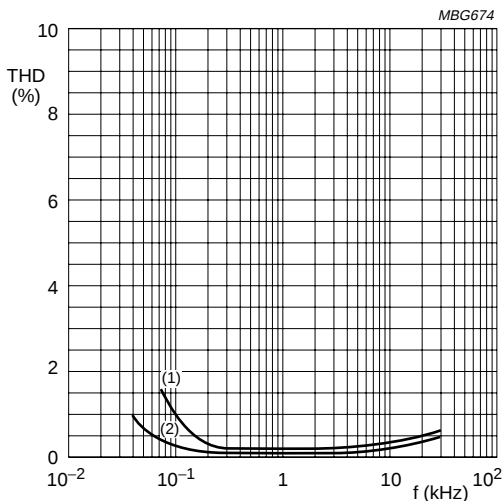


Fig.3 Quiescent current as a function of supply voltage.



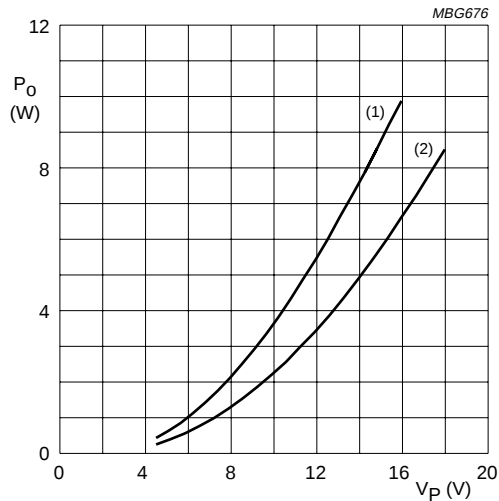
- (1) $R_L = 16 \Omega$.
- (2) $R_L = 8 \Omega$.

Fig.4 THD as a function of output power.



- (1) $G_v = 40$ dB; $P_o = 0.5$ W.
- (2) $G_v = 30$ dB; $P_o = 0.5$ W.

Fig.5 THD as a function of frequency.

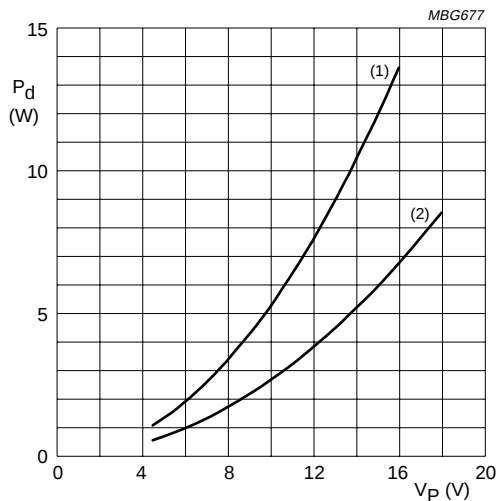


- THD = 10%; $f = 1$ kHz.
- (1) $R_L = 8 \Omega$.
 - (2) $R_L = 16 \Omega$.

Fig.6 Output power as a function of supply voltage.

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- (1) $R_L = 8 \Omega$.
- (2) $R_L = 16 \Omega$.

Fig.7 Total worst case power dissipation as a function of supply voltage.

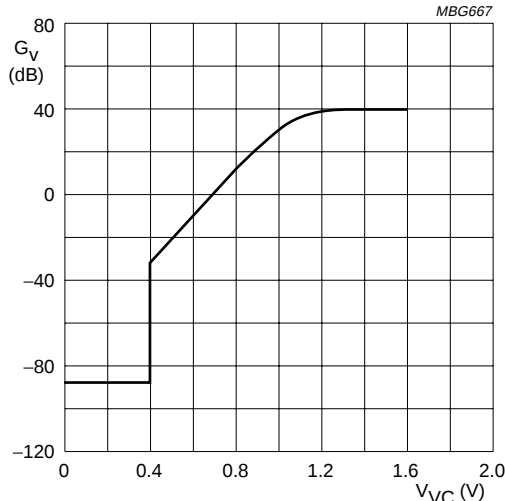
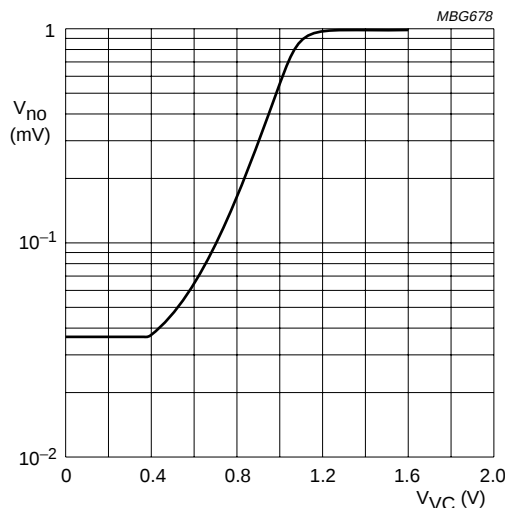
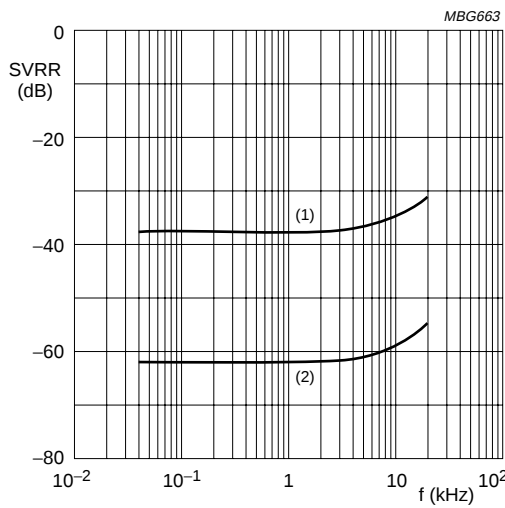


Fig.8 Voltage gain as a function of volume control voltage.



f = 22 Hz to 22 kHz.

Fig.9 Noise voltage as a function of volume control voltage.

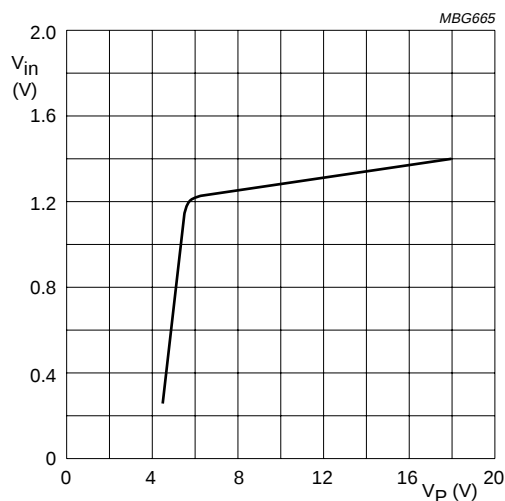


- (1) $V_{DC} = 1.4$ V; $V_{ripple} = 0.2$ V.
- (2) $V_{DC} = 0.4$ V; $V_{ripple} = 0.2$ V.

Fig.10 SVRR as a function of frequency.

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THD = 1 %.

Fig.11 Input signal handling.

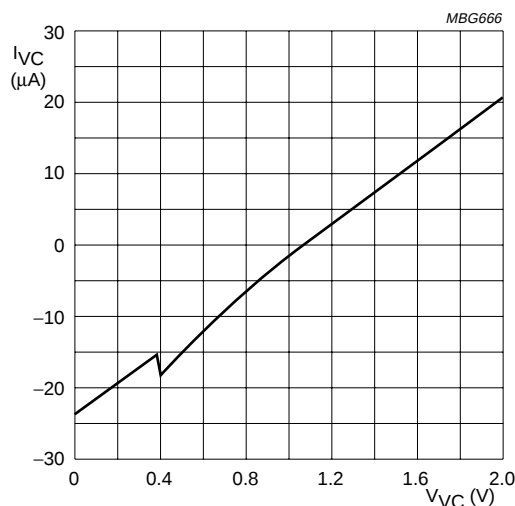


Fig.12 Volume control current as a function of volume control voltage.

APPLICATION INFORMATION

The application diagram is illustrated in Fig.13.

Test conditions

T_{amb} = 25 °C unless otherwise specified; V_P = 12 V;
V_{DC} = 1.4 V; f_i = 1 kHz; R_L = 16 Ω.

The quiescent current has been measured without load impedance.

The output power as a function of the supply voltage has been measured at THD = 10%. The maximum output power is limited by the maximum power dissipation and the maximum available output current.

The maximum input signal voltage is measured at THD = 1% at the output with a voltage gain of 0 dB.

To avoid instabilities and too high a distortion, the input ground and power ground must be separated as far as possible and connected as close as possible to the IC.

The DC volume control can be applied in several ways. Two possible circuits are shown below the main application diagram. The circuits at the control pin will influence the switch-on and switch-off behaviour and the maximum voltage gain.

For single-end applications the output peak current must not exceed 100 mA. At higher output currents the short-circuit protection (MCL) will be active.

Thermal considerations:

At high junction temperatures (>125 °C) the voltage gain will decrease when it is higher than 0 dB. This results in a decrease of the output voltage and an increase of the distortion level. Thus for an optimal performance of the IC the heatsink has to be designed properly.

Calculation example for application: V_P = 15 V; R_L = 8 Ω, stereo sine wave; worst case sine wave power dissipation is 12 W.

For T_{amb(max)} = 40 °C the thermal resistance from junction to ambient $R_{th\ j-a} = \frac{(125 - 40)}{12} = 7.1 \text{ K/W}$

The thermal resistance of the heatsink becomes:

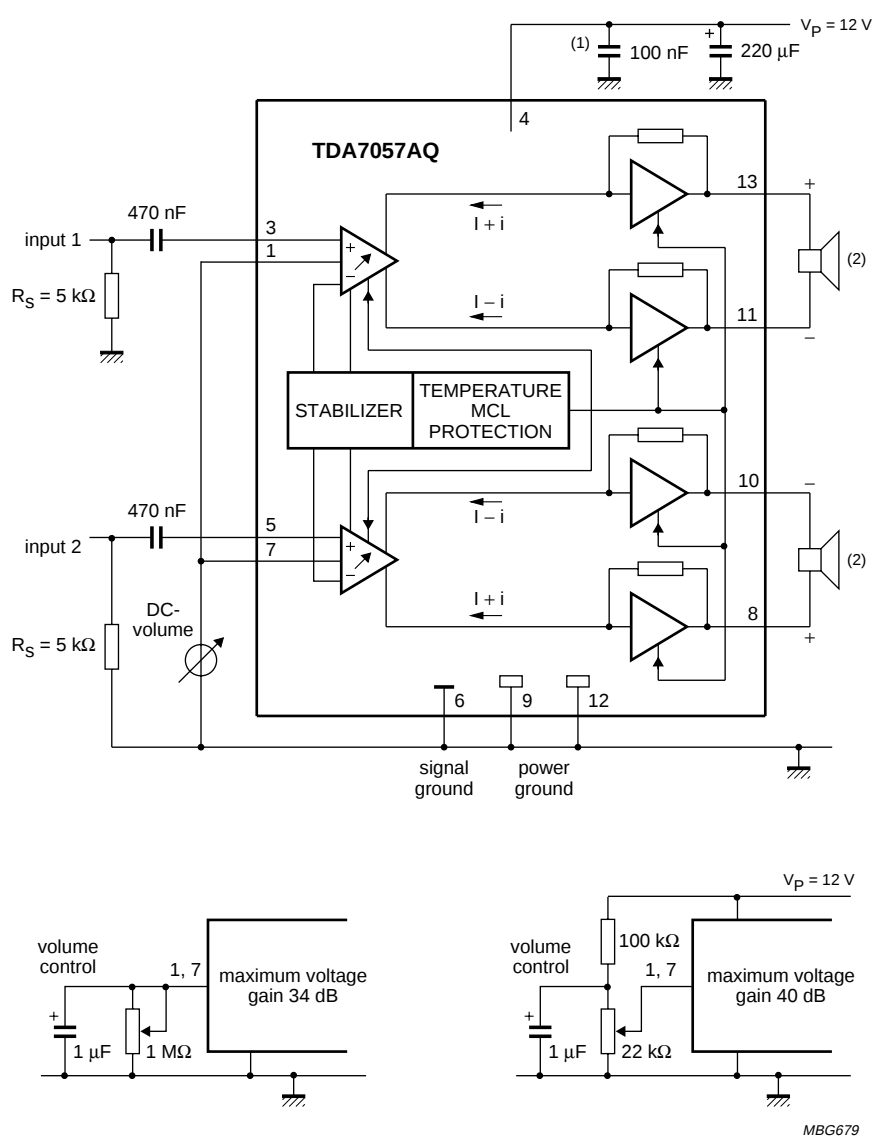
$$R_{th\ h-a} = R_{th\ j-a} - (R_{th\ j-c} + R_{th\ c-h});$$

$$R_{th\ h-a} = 7.1 - (4 + 0.1) = 3 \text{ K/W}.$$

It should be noted that for 'music power' the power dissipation will be approximately half of the sine wave dissipation. Thus a smaller heatsink can be used.

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- (1) This capacitor can be omitted if the 220 μF electrolytic capacitor is connected close to pin 5.
 (2) $R_L = 16 \Omega$.

Fig.13 Test and application diagram.

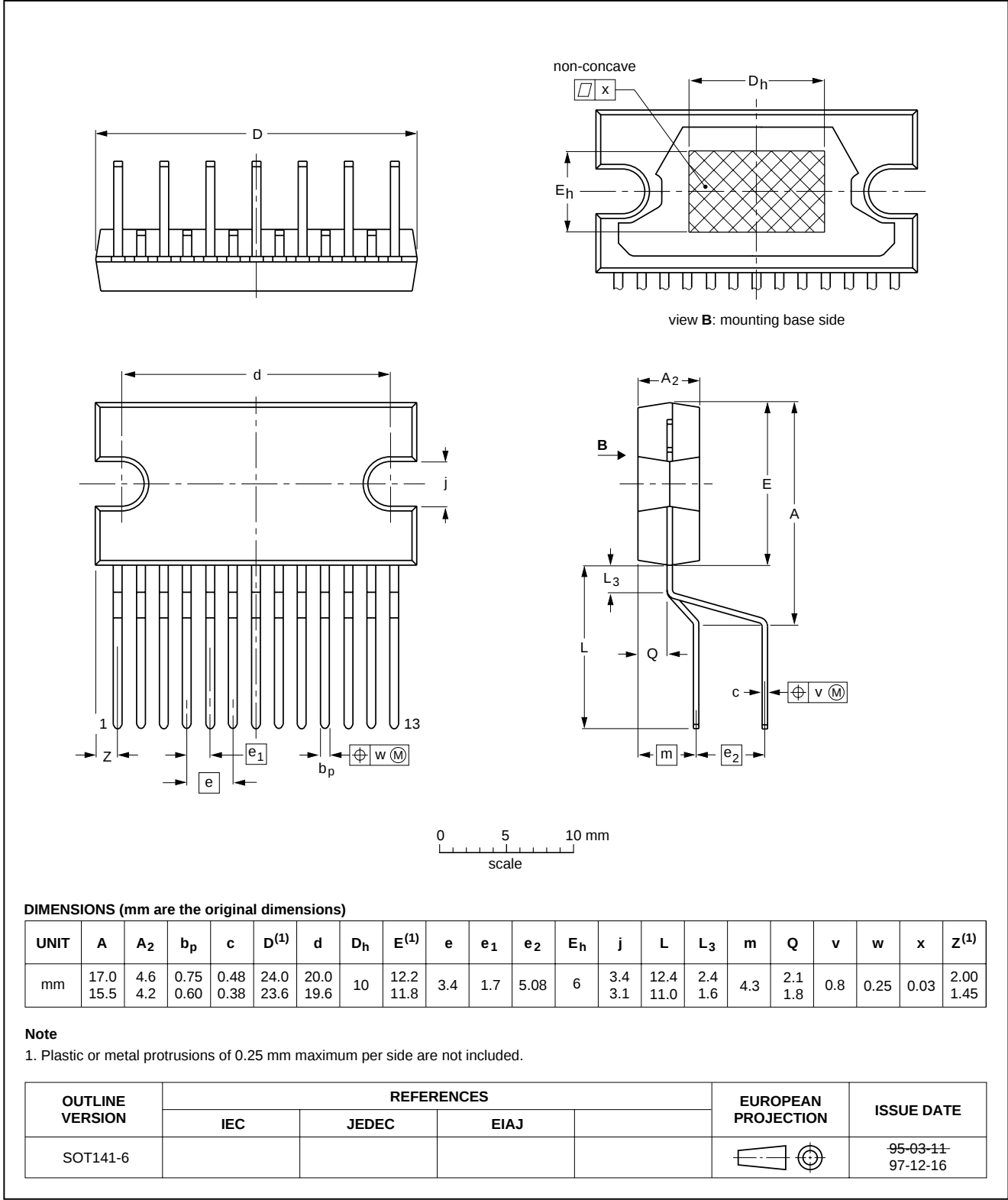
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PACKAGE OUTLINE

DBS13P: plastic DIL-bent-SIL power package; 13 leads (lead length 12 mm)

SOT141-6



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SOLDERING

Introduction

There is no soldering method that is ideal for all IC packages. Wave soldering is often preferred when through-hole and surface mounted components are mixed on one printed-circuit board. However, wave soldering is not always suitable for surface mounted ICs, or for printed-circuits with high population densities. In these situations reflow soldering is often used.

This text gives a very brief insight to a complex technology. A more in-depth account of soldering ICs can be found in our *"IC Package Databook"* (order code 9398 652 90011).

Soldering by dipping or by wave

The maximum permissible temperature of the solder is 260 °C; solder at this temperature must not be in contact with the joint for more than 5 seconds. The total contact time of successive solder waves must not exceed 5 seconds.

The device may be mounted up to the seating plane, but the temperature of the plastic body must not exceed the specified maximum storage temperature ($T_{\text{stg max}}$). If the printed-circuit board has been pre-heated, forced cooling may be necessary immediately after soldering to keep the temperature within the permissible limit.

Repairing soldered joints

Apply a low voltage soldering iron (less than 24 V) to the lead(s) of the package, below the seating plane or not more than 2 mm above it. If the temperature of the soldering iron bit is less than 300 °C it may remain in contact for up to 10 seconds. If the bit temperature is between 300 and 400 °C, contact may be up to 5 seconds.

DEFINITIONS

Data sheet status	
Objective specification	This data sheet contains target or goal specifications for product development.
Preliminary specification	This data sheet contains preliminary data; supplementary data may be published later.
Product specification	This data sheet contains final product specifications.
Limiting values	
Limiting values given are in accordance with the Absolute Maximum Rating System (IEC 134). Stress above one or more of the limiting values may cause permanent damage to the device. These are stress ratings only and operation of the device at these or at any other conditions above those given in the Characteristics sections of the specification is not implied. Exposure to limiting values for extended periods may affect device reliability.	
Application information	
Where application information is given, it is advisory and does not form part of the specification.	

LIFE SUPPORT APPLICATIONS

These products are not designed for use in life support appliances, devices, or systems where malfunction of these products can reasonably be expected to result in personal injury. Philips customers using or selling these products for use in such applications do so at their own risk and agree to fully indemnify Philips for any damages resulting from such improper use or sale.

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NOTES

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